

# BEM-Based Estimation for Time-Varying Channels and Training Design in Two-Way Relay Networks

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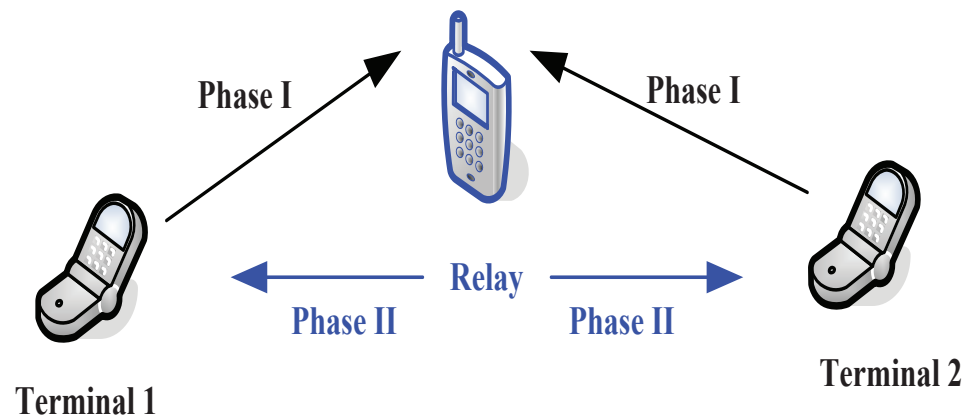
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# Outline

- Introduction
- Problem Formulation
- Proposed Solution
- Simulation Results
- Conclusion

## Introduction

- Two-way transmission was firstly exploited by Shannon [Shannon, 1961].
- Two-way relay networks (TWRN) now has drawn much attention due to its improved spectral efficiency over one-way relay networks (OWRN).
- The overall communication rate between two source terminals in TWRN is approximately twice that achieved in OWRN [Rankov, 2006].



# Problem Formulation

- For TWRNs, the relay and the two sources can be mobile, and the relative motion between any two nodes doubles the Doppler spread.
- This fact places additional demands on the estimation of time-varying channels.
- How to estimate time-varying channels  $f[n]$  and  $g[n]$ ? How to design training sequences ?

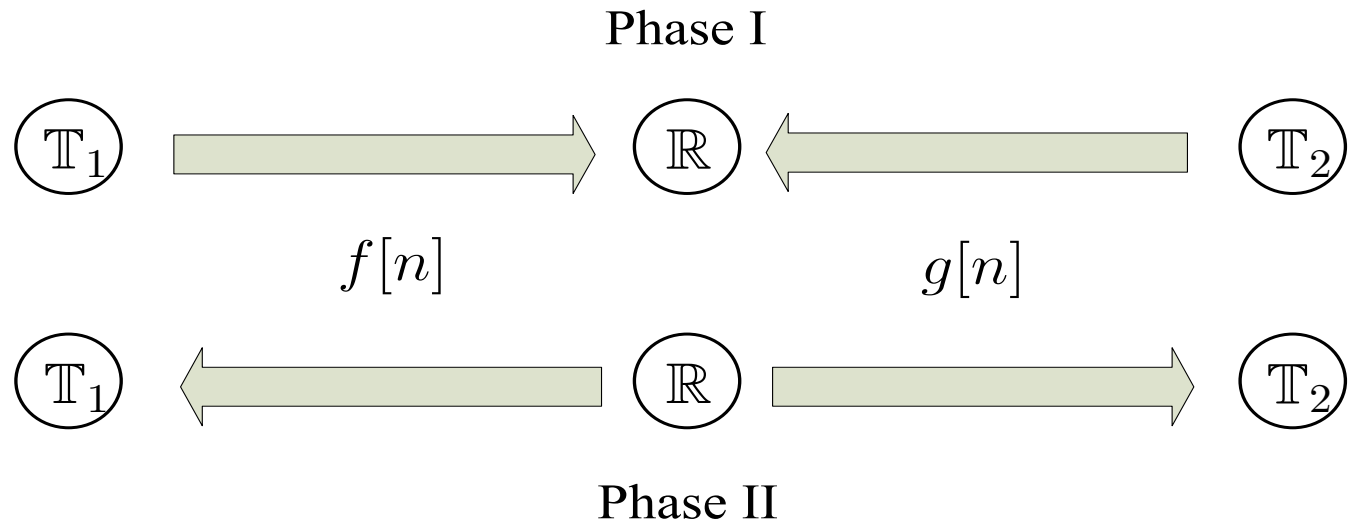


Figure 2: System model for TWRN with time varying channel.

# BEM + TLS + Optimal Training Sequence

- The BEM of  $f[n]$  and  $g[n]$  can be expressed as

$$f[n] = \sum_{q=0}^Q f_q e^{jw_q n}, \quad 0 \leq n \leq N - 1 \quad (1)$$

$$g[n] = \sum_{q=0}^Q g_q e^{jw_q n}, \quad 0 \leq n \leq N - 1. \quad (2)$$

- The received signal at  $\mathbb{T}_1$  can be written as

$$\begin{aligned} y[n + M + \Delta] = & \alpha \left( \sum_{q=0}^Q f_q e^{jw_q(n+M+\Delta)} \right) \left( \sum_{q=0}^Q f_q e^{jw_q n} \right) s_1[n] \\ & + \alpha \left( \sum_{q=0}^Q f_q e^{jw_q(n+M+\Delta)} \right) \left( \sum_{q=0}^Q g_q e^{jw_q n} \right) s_2[n] \\ & + \alpha f[n + M + \Delta] w_r[n] + w_1[n + M + \Delta]. \end{aligned} \quad (3)$$

## BEM

- Define  $\mathbf{A}$  as a  $M \times (Q + 1)^2$  Vandermonde matrix with the  $(i(Q + 1) + k)$ th column

$$\mathbf{A}[:, i(Q + 1) + k] = e^{jw_i(M+\Delta)} \begin{bmatrix} 1 \\ e^{j(w_i+w_k)} \\ \vdots \\ e^{j(M-1)(w_i+w_k)} \end{bmatrix}, \quad (4)$$

$$i, k = 0, \dots, Q,$$

- Then we can obtain

$$\mathbf{y} = \alpha \mathbf{S}_1 \mathbf{A} \mathbf{h}_1 + \alpha \mathbf{S}_2 \mathbf{A} \mathbf{h}_2 + \underbrace{\alpha \mathbf{f} \odot \mathbf{w}_r + \mathbf{w}_1}_{\mathbf{w}}, \quad (5)$$

where  $\odot$  denotes the Hadamard product,  $\mathbf{h}_1$  is the  $(Q + 1)^2 \times 1$  vector with the  $(i(Q + 1) + k)$ th entry  $f_i f_k$ ,  $\mathbf{h}_2$  is the  $(Q + 1)^2 \times 1$  vector with the  $(i(Q + 1) + k)$ th entry  $f_i g_k$ , and  $\mathbf{w}$  denotes the equivalent combined noise.

## BEM (continued)

- Lemma: Matrix  $\mathbf{A}$  defined in (4) has rank  $2Q + 1$ .
- Combining those linearly dependent columns in  $\mathbf{A}$ , we can rewrite (5) as

$$\mathbf{y} = \alpha \mathbf{S}_1 \mathbf{A}_0 \mathbf{x}_1 + \alpha \mathbf{S}_2 \mathbf{A}_0 \mathbf{x}_2 + \mathbf{w}, \quad (6)$$

where  $\mathbf{x}_i$  is a  $(2Q + 1) \times 1$  vector with entries

$$\mathbf{x}_1(m) = \sum_{i+k=m} e^{jw_i(M+\Delta)} f_i f_k, \quad m = 0, \dots, 2Q \quad (7)$$

$$\mathbf{x}_2(m) = \sum_{i+k=m} e^{jw_i(M+\Delta)} f_i g_k, \quad (8)$$

$\mathbf{A}_0$  is an  $M \times (2Q + 1)$  Vandermonde matrix defined as

$$\begin{bmatrix} 1 & 1 & \dots & 1 & 1 \\ e^{j2w_0} & e^{j(w_0+w_1)} & \dots & e^{j(w_{Q-1}+w_Q)} & e^{j2w_Q} \\ \vdots & \vdots & \dots & \vdots & \vdots \\ e^{j2w_0(M-1)} & e^{j(w_0+w_1)(M-1)} & \dots & e^{j(w_{Q-1}+w_Q)(M-1)} & e^{j2w_Q(M-1)} \end{bmatrix}$$

# Total Least Square (TLS) Method

- Equation (6) can be rewritten as

$$\mathbf{y} = \underbrace{[\alpha\mathbf{S}_1\mathbf{A}_0, \alpha\mathbf{S}_2\mathbf{A}_0]}_{\mathbf{C}} \underbrace{\begin{bmatrix} \mathbf{x}_1 \\ \mathbf{x}_2 \end{bmatrix}}_{\mathbf{x}} + \mathbf{w}, \quad (9)$$

- By assuming that  $\mathbf{y}$  and  $\mathbf{C}$  contain disturbance  $\mathbf{e}$  and  $\mathbf{E}$ , respectively, we obtain:

$$(\mathbf{C} + \mathbf{E})\mathbf{x} = \mathbf{y} + \mathbf{e}, \Leftrightarrow \underbrace{[-\mathbf{y}, \mathbf{C}]}_{\mathbf{J}} \mathbf{z} + \underbrace{[-\mathbf{e}, \mathbf{E}]}_{\mathbf{D}} \mathbf{z} = \mathbf{0}_{M \times 1}, \quad (10)$$

where  $\mathbf{z} = [1, \mathbf{x}^T]^T$ .

- The TLS finds  $\mathbf{x}$ ,  $\mathbf{E}$ , and  $\mathbf{e}$  from the following optimization:

$$\min_{\mathbf{D}, \mathbf{x}} \|\mathbf{D}\mathbf{z}\|_F^2, \quad (11)$$

$$s.t. \quad (\mathbf{y} + \mathbf{e}) \in \text{Range}(\mathbf{C} + \mathbf{E}), \quad (12)$$

where  $\|\cdot\|_F^2$  denotes the Frobenius norm of a matrix.



# Total Least Square (TLS) Method

- We solve the following optimization:

$$\min_{\mathbf{z}} \|\mathbf{J}\mathbf{z}\|_F^2, \quad (13)$$

- The solution to (13) is easily proved to be eigenvector of  $\mathbf{J}^H \mathbf{J}$  that corresponds to the smallest eigenvalue.
- The estimation of  $\mathbf{x}$  can be expressed as

$$\hat{\mathbf{x}} = \frac{1}{\mathbf{V}(1, 4Q + 3)} \begin{bmatrix} \mathbf{V}(2, 4Q + 3) \\ \vdots \\ \mathbf{V}(4Q + 3, 4Q + 3) \end{bmatrix}, \quad (14)$$

where  $\mathbf{V}(m_1, m_2)$  is the  $(m_1, m_2)$ th entry of  $\mathbf{V}$ .

$$\mathbf{J} = \mathbf{U} \begin{bmatrix} \sigma_1(\mathbf{J}) & & & 0 \\ & \sigma_2(\mathbf{J}) & & \\ & & \ddots & \\ 0 & & & \sigma_{4Q+3}(\mathbf{J}) \end{bmatrix} \mathbf{V}^H, \quad (15)$$

# Training Sequence Design

- Consider the optimization problem

$$\arg \min_{\mathbf{s}_1, \mathbf{s}_2} \kappa(\mathbf{C}(\mathbf{s}_1, \mathbf{s}_2)) = \frac{\sigma_1(\mathbf{C})}{\sigma_{4Q+2}(\mathbf{C})}, \quad (16)$$

where  $\kappa(\cdot)$  denotes the condition number.

- **Type-1 Training:** The training symbol from  $\mathbb{T}_1$  and  $\mathbb{T}_2$  appear alternatively with maximum power  $P_s$  and arbitrary phase.
- **Type-2 Training:** We choose  $|s_1[n]| = P_s$  with arbitrary phase and set  $s_2[n] = (-1)^n s_1[n]$ .
- **Theorem:** Denote  $\mathbf{C}$  from type-1 training as  $\mathbf{C}_1$ , and that from type-2 training as  $\mathbf{C}_2$ . The condition numbers of  $\mathbf{C}_1$  and  $\mathbf{C}_2$  are the same.

# Simulation Results

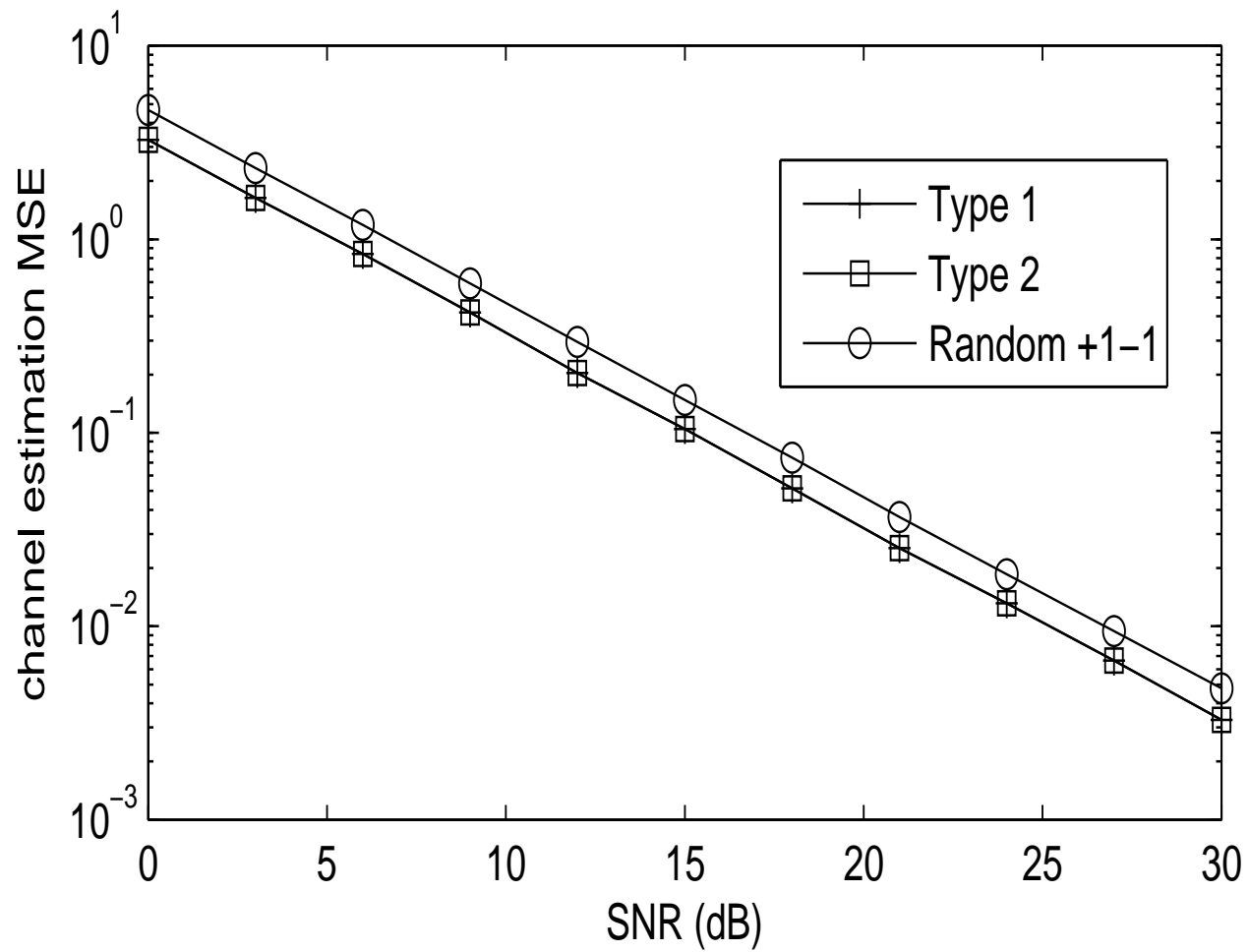


Figure 3: Channel estimation MSE of  $\mathbf{x}$

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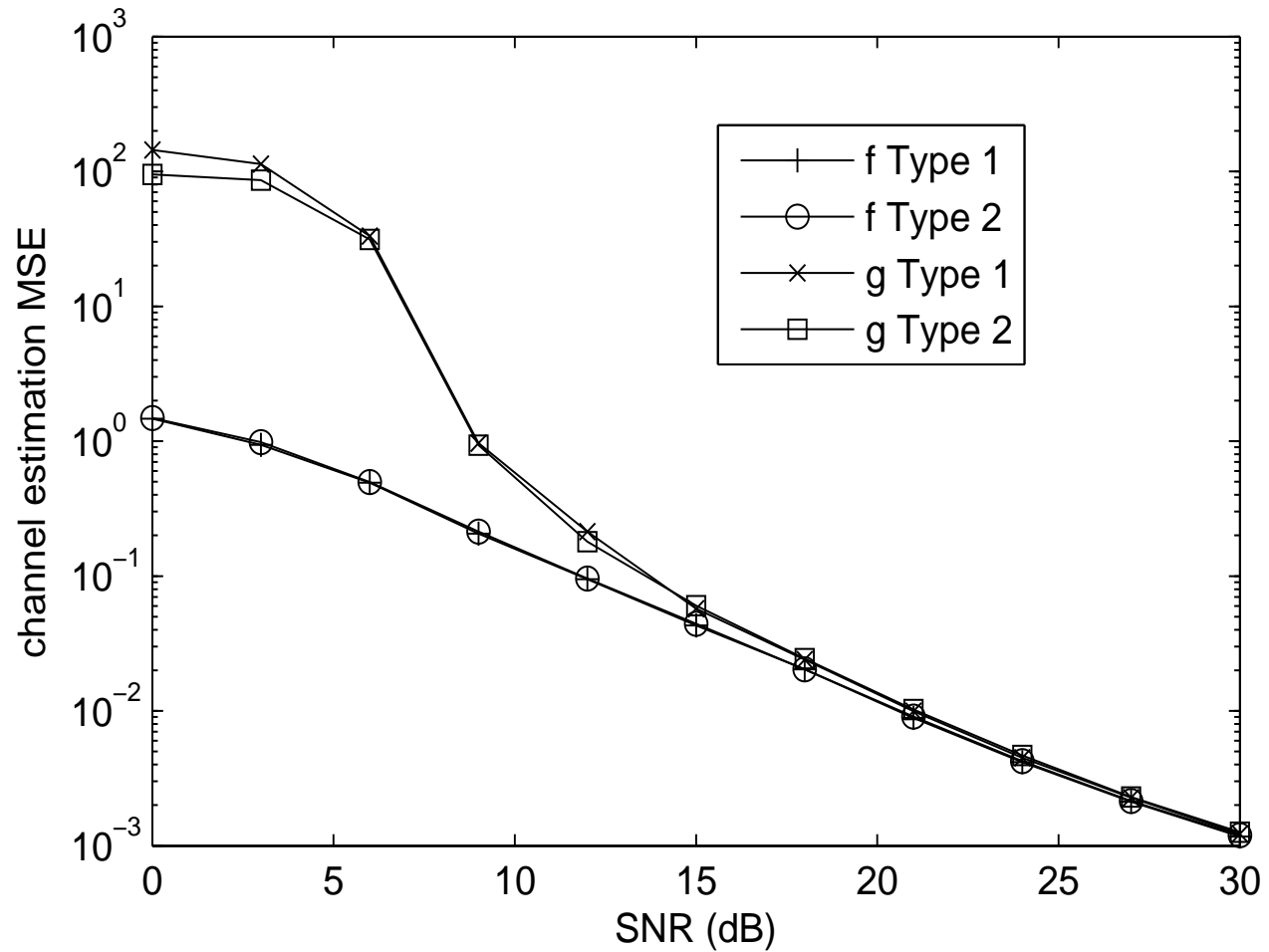


Figure 4: Channel estimation MSE of f and g

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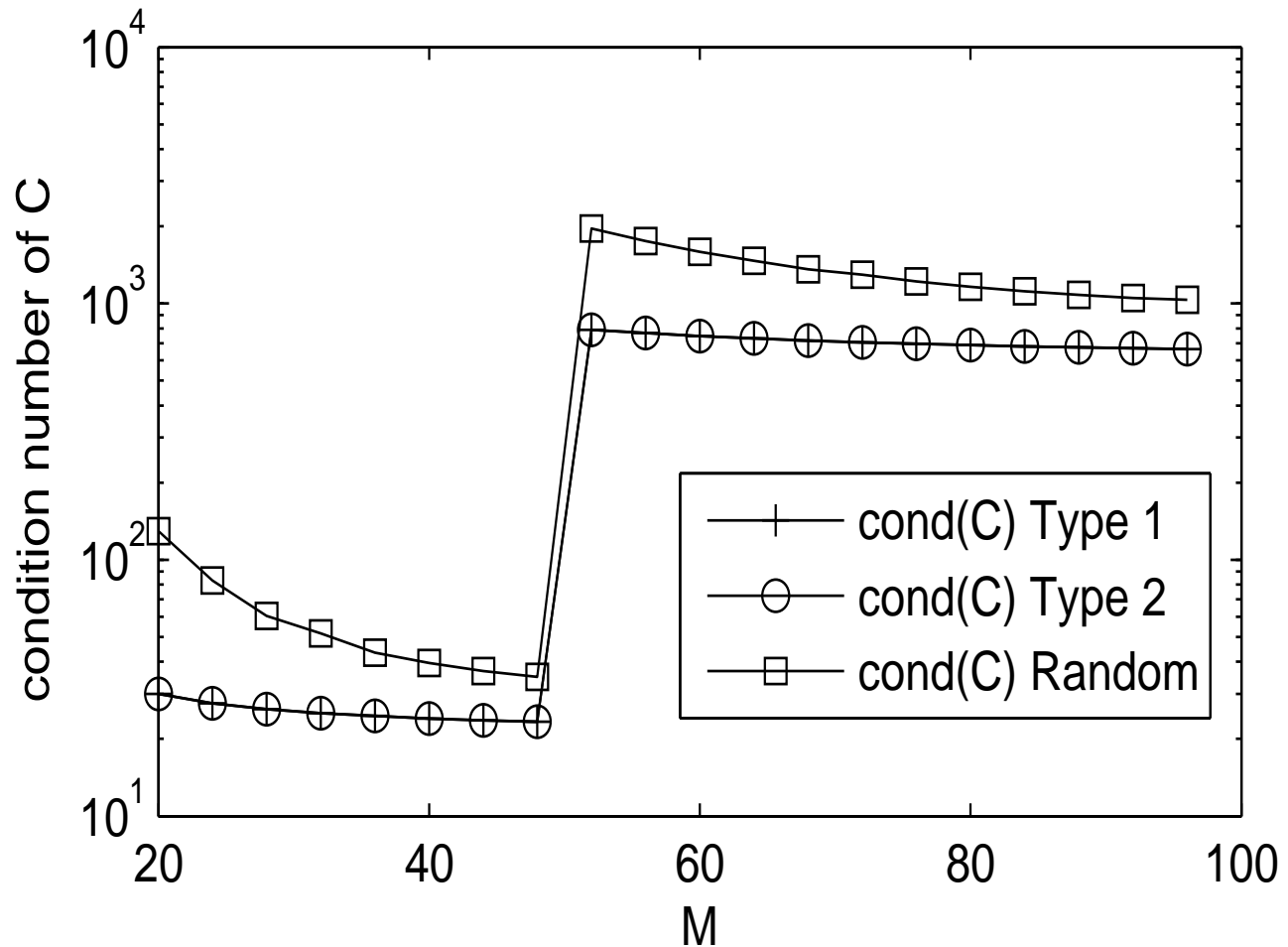


Figure 5: Condition number of  $A_0$

# Conclusion

## ■ Our contribution:

1. Time-varying channel estimation in AF TWRN
2. BEM + TLS
3. Training sequence design.

## ■ **Problem:** How to improve channel estimation performance ?

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## ■ ICC 2011